



POSTAL BOOK PACKAGE 2026

ELECTRONICS ENGINEERING

.....

CONVENTIONAL Practice Sets

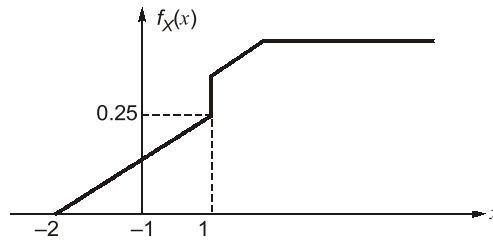
CONTENTS

COMMUNICATION SYSTEMS

| | |
|--|----------|
| 1. Theory of Random Variable and Noise | 2 - 21 |
| 2. Amplitude Modulation | 22 - 34 |
| 3. Angle Modulation | 35 - 49 |
| 4. AM Transmitters and Receivers | 50 - 56 |
| 5. Pulse Modulation | 57 - 69 |
| 6. Data Transmission Schemes | 70 - 75 |
| 7. Optimum Receivers for AWGN Channels | 76 - 89 |
| 8. Information Theory and Coding | 90 - 102 |

Theory of Random Variable and Noise

Q1 Define PDF and summarise its important properties. Also calculate the probability of outcome of a Random Variable (RV) X having $X \leq 1$ for the following PDF curve of RV as shown.



Solution:

Probability density function specifies the probability of a random variable taking a particular value.

The Probability Density Function (PDF) which is generally denoted by $f_X(x)$ or $P_X(x)$ or $p_X(x)$ is defined in terms of the Cumulative Distribution Function (CDF) $F_X(x)$ as,

$$\text{PDF} = f_X(x) = \frac{d}{dx} F_X(x) \quad \dots(i)$$

The PDF has the following properties:

- (i) $f_X(x) \geq 0$ for all x

This results from the fact that probability cannot be negative. Also, $F_X(x)$ increases monotonically, as x increases, more outcomes are included in the prob. of occurrence represented by $F_X(x)$.

- (ii) Area under the PDF curve is always equal to unity.

i.e.
$$\int_{-\infty}^{\infty} f_X(x) dx = 1$$

- (iii) The CDF is obtained by the result

$$\text{CDF} = \int_{-\infty}^x f_X(x) dx$$

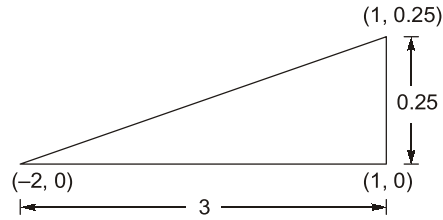
- (iv) Probability of occurrence of the value of random variable between the limits of x_1 and x_2 is given by,

$$P(x_1 < X < x_2) = \int_{x_1}^{x_2} f_X(x) dx$$

Now consider the given PDF curve, since we have to find $P(x \leq 1)$ so,

Equation for the PDF curve for $x \leq 1$ is,

$$f_X(x) = \left(\frac{1}{12}x + \frac{1}{6} \right)$$



Now, $P(x \leq 1)$

$$= P(-2 < x < 1) = \int_{-2}^1 \left(\frac{1}{12}x + \frac{1}{6} \right) dx = \left[\frac{1}{12} \cdot \frac{x^2}{2} + \frac{1}{6}x \right]_{-2}^1 = \frac{3}{8}$$

$$\therefore P(x \leq 1) = \frac{3}{8}$$

Q2 Find the cumulative distribution function $F(x)$ corresponding to the PDF $f(x) = \frac{1}{\pi(1+x^2)}$, $-\infty < x < \infty$.

Solution:

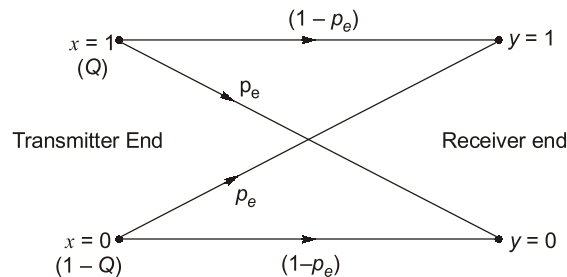
Given

$$f(x) = \frac{1}{\pi(1+x^2)}, -\infty < x < \infty$$

$$F(x) = P(X \leq x)$$

$$= \int_{-\infty}^x f(x) dx = \frac{1}{\pi} \int_{-\infty}^x \frac{dx}{1+x^2} = \frac{1}{\pi} [\tan^{-1} x]_{-\infty}^x = \frac{1}{\pi} \left(\frac{\pi}{2} + \tan^{-1} x \right)$$

Q3 A BSC (Binary Symmetric Channel) error probability is P_e . The probability of transmitting '1' is Q , and that of transmitting '0' is $(1-Q)$ as in figure below. Calculate the probabilities of receiving 1 and 0 at the receiver?



Solution:

If x and y are the transmitted digit and the received digit respectively, then for a BSC,

$$P_{y|x}(0|1) = P_{y|x}(1|0) = P_e$$

$$P_{y|x}(0|0) = P_{y|x}(1|1) = 1 - P_e$$

Also,

$$P_x(1) = Q \text{ and } P_x(0) = 1 - Q$$

We have to find, $P_y(1)$ and $P_y(0)$ = ?

\therefore

$$P_y(1) = P_x(0) P_{y|x}(1|0) + P_x(1) P_{y|x}(1|1) = (1-Q)P_e + Q(1-P_e)$$

also,

$$P_y(0) = P_x(0) P_{y|x}(0|0) + P_x(1) P_{y|x}(0|1) = (1-Q)(1-P_e) + QP_e$$

Q4 For the triangular distribution

$$f(x) = \begin{cases} x, & 0 \leq x \leq 1 \\ 2-x, & 1 \leq x \leq 2 \\ 0, & \text{otherwise} \end{cases}$$

Find the mean and variance.

Solution:

$$\begin{aligned}
 \text{Mean} = E(X) &= \int_{-\infty}^{\infty} xf(x)dx = \int_0^1 x \cdot x dx + \int_1^2 x(2-x)dx = \int_0^1 x^2 dx + \int_1^2 (2x - x^2)dx \\
 &= \left[\frac{x^3}{3} \right]_0^1 + \left[2\left(\frac{x^2}{2}\right) - \frac{x^3}{3} \right]_1^2 \\
 &= \frac{1}{3} + \left[\left(4 - \frac{8}{3}\right) - \left(1 - \frac{1}{3}\right) \right] = \frac{1}{3} + \frac{4}{3} - \frac{2}{3} = 1 \\
 E(X^2) &= \int_{-\infty}^{\infty} x^2 f(x)dx = \int_0^1 x^2 x dx + \int_1^2 x^2 (2-x)dx \\
 &= \int_0^1 x^3 dx + \int_1^2 (2x^2 - x^3)dx = \left[\frac{x^4}{4} \right]_0^1 + \left[2\left(\frac{x^3}{3}\right) - \frac{x^4}{4} \right]_1^2 \\
 &= \frac{1}{4} + \left[\left(\frac{16}{3} - \frac{16}{4}\right) - \left(\frac{2}{3} - \frac{1}{4}\right) \right] = \frac{1}{4} + \frac{16}{3} - 4 - \frac{2}{3} + \frac{1}{4} = \frac{7}{6} \\
 \text{Var}(X) &= E(X^2) - E(X)^2 = \frac{7}{6} - (1)^2 = \frac{1}{6}
 \end{aligned}$$

Q5 The joint density function of two continuous random variables is given by

$$f(x, y) = \begin{cases} xy/8, & 0 < x < 2, 1 < y < 3 \\ 0, & \text{otherwise} \end{cases}$$

Find (a) $E(X)$, (b) $E(Y)$ and (c) $E(2X + 2Y)$.**Solution:**

$$\begin{aligned}
 \text{(a)} \quad E(X) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} xf(x, y)dx dy = \int_{x=0}^2 \int_{y=1}^3 x(xy/8)dx dy = \frac{4}{3} \\
 \text{(b)} \quad E(Y) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} yf(x, y)dx dy = \int_{x=0}^2 \int_{y=1}^3 y(xy/8)dx dy = \frac{13}{6} \\
 \text{(c)} \quad E(2X + 3Y) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} (2x + 3y)dx dy = \int_{x=0}^2 \int_{y=1}^3 (2x + 3y)(xy/8)dx dy = \frac{55}{6}
 \end{aligned}$$

Q6 A WSS random process $x(t)$ is applied to the input of an LTI system with impulse response

$$h(t) = 3e^{-2t} u(t)$$

Find the mean value of the output $y(t)$ of the system, if $E[x(t)] = 2$. Here $E[\cdot]$ denotes the expectation operator.**Solution:**The output $y(t)$ is the convolution of the input $x(t)$ and the impulse response $h(t)$.

$$y(t) = \int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) \cdot d\tau$$

 \therefore

$$E[y(t)] = \int_{-\infty}^{\infty} h(\tau) \cdot E[x(t - \tau)] \cdot d\tau$$

$$E[y(t)] = H(0) \times E[x(t)]$$

$$E[y(t)] = E[x(t)] \cdot H(0)$$

where, $H(0) = H(\omega)|_{\omega=0}$ and $H(\omega)$ = Fourier transform of $h(t)$

Given $E[x(t)] = 2$, $h(t) = 3e^{-2t} u(t)$

Taking Fourier transform, $H(\omega) = \frac{3}{2 + j\omega} \Rightarrow H(0) = \frac{3}{2}$

$$E[y(t)] = 2 \times \frac{3}{2} = 3$$

Q.7 Suppose that two signals $s_1(t)$ and $s_2(t)$ are orthogonal over the interval $(0, T)$. A sample function $n(t)$ of a zero-mean white noise process is correlated with $s_1(t)$ and $s_2(t)$ separately, to yield the following variables:

$$n_1 = \int_0^T s_1(t) n(t) dt \quad \text{and} \quad n_2 = \int_0^T s_2(t) n(t) dt$$

Prove that n_1 and n_2 are orthogonal.

Solution:

$$\begin{aligned} E[n_1 n_2] &= E \left[\int_0^T s_1(u) n(u) du \int_0^T s_2(v) n(v) dv \right] \\ &= \int_0^T \int_0^T s_1(u) s_2(v) E[n(u) n(v)] du dv \end{aligned}$$

$n(t)$ is a white noise process.

So, $R_N(\tau) = \frac{N_0}{2} \delta(\tau)$

$$E[n(u) n(v)] = \frac{N_0}{2} \delta(u - v)$$

Hence,
$$\begin{aligned} E[n_1 n_2] &= \frac{N_0}{2} \int_0^T \int_0^T s_1(u) s_2(v) \delta(u - v) du dv \\ &= \frac{N_0}{2} \int_0^T s_1(u) s_2(u) du \\ &= 0 \quad \because s_1(t) \text{ and } s_2(t) \text{ are orthogonal over } (0, T) \end{aligned}$$

$E[n_1 n_2] = 0$. So, n_1 and n_2 are also orthogonal.

Q.8 Find the time autocorrelation function of the signal $g(t) = e^{-at} u(t)$ and from this obtain the energy spectral density (ESD) of $g(t)$.

Solution:

Auto correlation function,

$$\begin{aligned} R_x(\tau) &= \int_{-\infty}^{\infty} g(t) g(t - \tau) dt = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-a(t-\tau)} u(t - \tau) dt \\ &= \int_{\tau}^{\infty} e^{-2at} e^{a\tau} dt = e^{a\tau} \int_{\tau}^{\infty} e^{-2at} dt = \frac{e^{a\tau}}{-2a} \left[e^{-2at} \right]_{\tau}^{\infty} = \frac{e^{a\tau}}{2a} e^{-2a\tau} = \frac{e^{-a\tau}}{2a} \end{aligned}$$

Similar process is valid for negative side because for real $g(t)$, $R_x(\tau)$ is even function of time

$$R_x(\tau) = \frac{e^{-a|\tau|}}{2a}$$

Now we know that

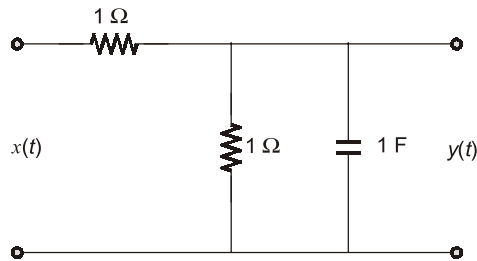
$$R_x(\tau) \xrightarrow{F.T} S_x(\omega)$$

Energy spectral density

$$S_x(\omega) = \frac{1}{\omega^2 + a^2} \triangleq |G(\omega)|^2, \text{ where } G(\omega) = F.T[g(t)]$$

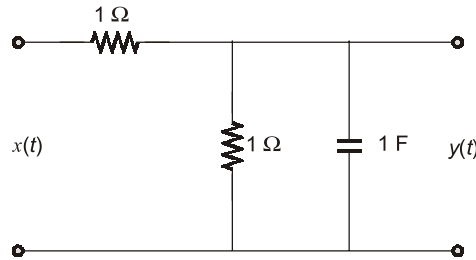
Q9 If the input to a low-pass filter as shown below in Figure is a random process $x(t)$ with autocorrelation function, $R_x(\tau) = 5\delta(\tau)$, then find

- Power spectral density of the output random process;
- Average power of output random process.



Solution:

Given low pass filter,



By taking Laplace transform

$$\text{Transfer function, } H(s) = \frac{1}{s+2}; \quad H(j\omega) = \frac{1}{j\omega+2}$$

Input Auto-correlation function

$$R_x(\tau) = 5\delta(\tau)$$

$$\text{Power spectral density } S_x(\omega) = F[R_x(\tau)] = 5$$

Output power spectral density

$$S_y(\omega) = |H(j\omega)|^2 S_x(\omega)$$

$$S_y(\omega) = \frac{5}{\omega^2 + 4}$$

$$\text{Output Auto-correlation function} = R_y(\tau) = \frac{5}{4} e^{-2|\tau|}$$

\therefore

$$e^{-a|\tau|} \xleftrightarrow{\text{CTFT}} \frac{2a}{\omega^2 + a^2}$$

(ii) Average power of the output process = $R_y(0)$

$$P_y = \frac{5}{4} \text{ W}$$

